

EE105 – Fall 2014

Microelectronic Devices and Circuits

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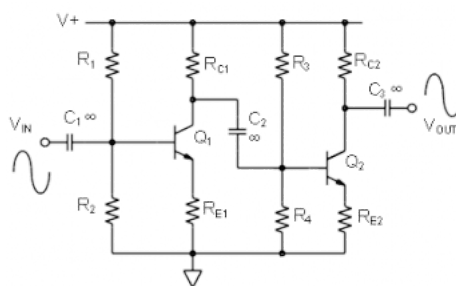


Lecture22-Multistage Amplifiers(2)

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AC-Coupled CE Amplifiers



- Discussed in Lecture 21
- Each stage has independent DC bias
- Cannot achieve very low frequency response
 - e.g., Lab 5 and 6 use DC coupling

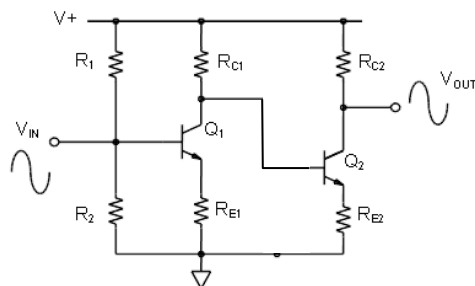


Lecture22-Multistage Amplifiers(2)

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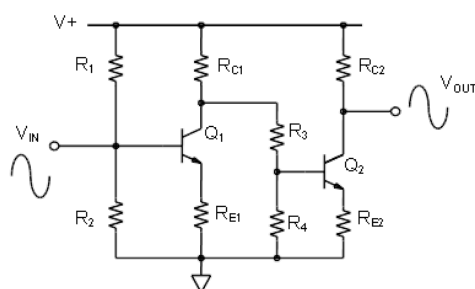
DC-Coupled Cascaded Common-Emitter Amplifiers



- In DC coupled multistage cascaded CE amplifiers, the output bias level of each stage increases to maintain the collector more positive than the base
 - If this voltage “stacking” is severe, little head room is left in the final stages of the cascade.



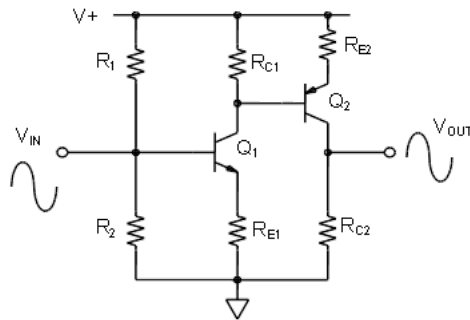
DC-Coupled Cascaded Common-Emitter Amplifiers



- The R_3, R_4 resistor divider not only reduces the signal amplitude seen at the base of Q_2 , it also reduces the DC bias level from the collector of Q_1 to a more manageable DC level at the base of Q_2 .
- This happens at the cost of overall signal gain in the combined amplifier.



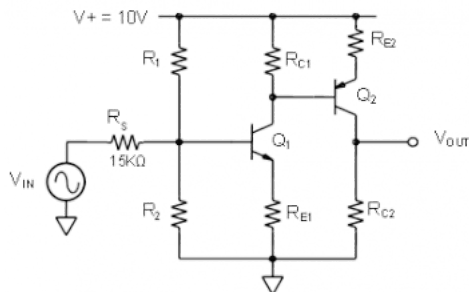
Complementary Pair Amplifier



- Having both polarities of transistors allows for more flexibility in how amplifier stages can be combined and can make biasing easier as well.
- First stage: NPN CE
Second stage: PNP CE
- By using complementary devices, active level shifting can be combined with amplification
 - The p-stage collector DC operating point tends to cancel the bias level “stacking” issue we encountered in the all NPN CE amplifier



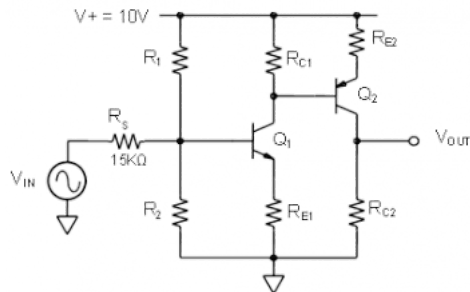
Complementary Pair Amplifier – Design Example



- Design a Complementary Pair amplifier stage using 2N3904 and 2N3906 transistors
 - $\beta \sim 120$, $V_{BE} \sim 0.7V$
 - $V_{CC} = 10V$
 - $R_S = 15 K\Omega$
- Question:
 - Estimate the DC bias voltages
 - Determine the small signal gain at 1 KHz



Complementary Pair Amplifier – Design Example



DC bias:

$$V_{EQ} = V_{CC} \frac{R_2}{R_1 + R_2}$$

$$R_B = R_1 \parallel R_2$$

$$I_{C1} \approx \frac{V_{EQ} - 0.7V}{R_B / \beta + R_{E1}}$$

$$V_{C1} \approx V_{CC} - I_{C1} R_{C1}$$

$$I_{C2} \approx \frac{V_{CC} - V_{C1} - 0.7V}{R_{E2}}$$

Small-signal gain:

Both Q_1 and Q_2 are CE with degeneration:

$$A_{V1} = -\frac{g_{m1} R_{C1}}{1 + g_{m1} R_{E1}} \approx -\frac{R_{C1}}{R_{E1}}$$

$$A_{V2} \approx -\frac{R_{C2}}{R_{E2}}$$

$$A_V = \frac{R_B}{R_S + R_B} \left(\frac{R_{C1}}{R_{E1}} \right) \left(\frac{R_{C2}}{R_{E2}} \right)$$

$$R_B = R_1 \parallel R_2 \parallel (r_\pi + (1 + \beta) R_{E1}) \approx R_1 \parallel R_2$$

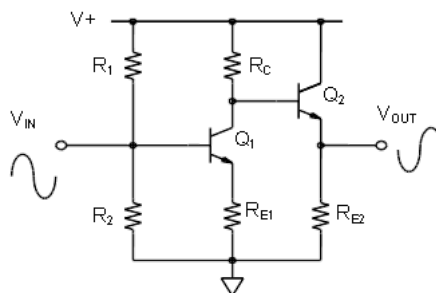


Lecture22-Multistage Amplifiers(2)

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DC-Coupled CE-CC Amplifier



$$I_{C1} = \frac{V_{EQ} - 0.7V}{R_B / \beta + R_{E1}}$$

$$V_{C1} = V_{CC} - I_{C1} R_C$$

$$I_{C2} = \frac{V_{C1} - 0.7V}{R_{E2}}$$

- Input stage: CE
 - High R_{in}
 - Moderate to high gain
- Output stage: CC
 - Low R_{out}
- Overall gain is nearly independent of load resistance and source resistance
- Multiple CE stages can be cascaded with CC stages inserted between them to reduce the attenuation due to inter-stage loading
- Direct coupling is commonly used in IC because this saves biasing resistors



Lecture22-Multistage Amplifiers(2)

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